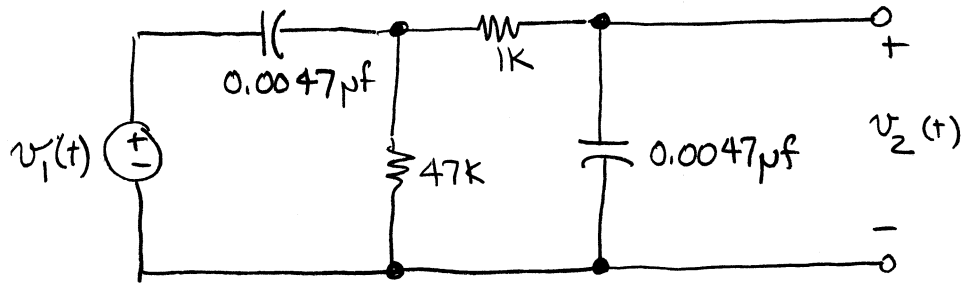
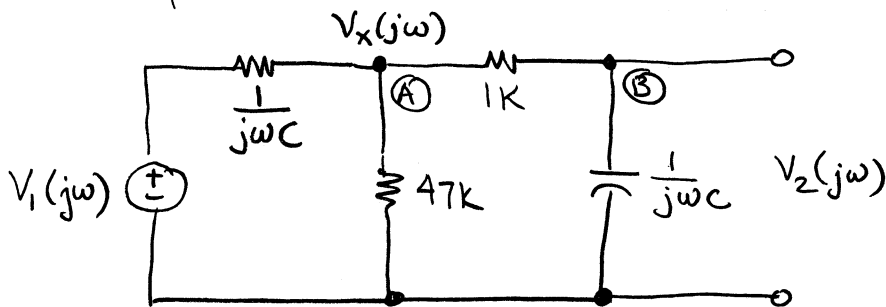


Analysis of directly coupled high-pass/low-pass filter cascade.

1



Convert to phasors.



Do KCL at nodes (A) and (B). Assume the voltage at node (A) is $V_x(j\omega)$.

$$\frac{V_1 - V_x}{\frac{1}{j\omega C}} - \frac{V_x - 0}{47k} - \frac{V_x - V_2}{1k} = 0 \quad (1)$$

$$\frac{V_x - V_2}{1k} - \frac{V_2 - 0}{\frac{1}{j\omega C}} = 0$$

$$\frac{V_x}{1k} - \frac{V_2}{1k} - j\omega C V_2 = 0$$

$$V_x = 1k \left(\frac{1}{1k} + j\omega C \right) V_2 = (1 + j\omega(47k)C) V_2 \quad (2)$$

Expanding (1) gives

$$j\omega C V_1 - j\omega C V_x - \frac{1}{47k} V_x - \frac{V_x}{1k} + \frac{V_2}{1k} = 0$$

Multiply through by $1k$

$$j\omega(1k)c V_1 - j\omega(1k)c V_x - \frac{1}{47} V_x - V_x + V_2 = 0$$

Collect terms

$$j\omega(1k)c V_1 - V_x \left[j\omega(1k)c + \frac{1}{47} + 1 \right] + V_2 = 0$$

↑
substituting from (2)

$$j\omega(1k)c V_1 - [1 + j\omega(1k)c] \left[1 + \frac{1}{47} + j\omega(1k)c \right] V_2 + V_2 = 0$$

Rearranging

$$T(j\omega) = \frac{V_2}{V_1} = \frac{j\omega(1k)c}{[1 + j\omega(1k)c] \left[1 + \frac{1}{47} + j\omega(1k)c \right] + 1}$$

To plot this I will make two approximations

(1) $1 + \frac{1}{47} \approx 1$

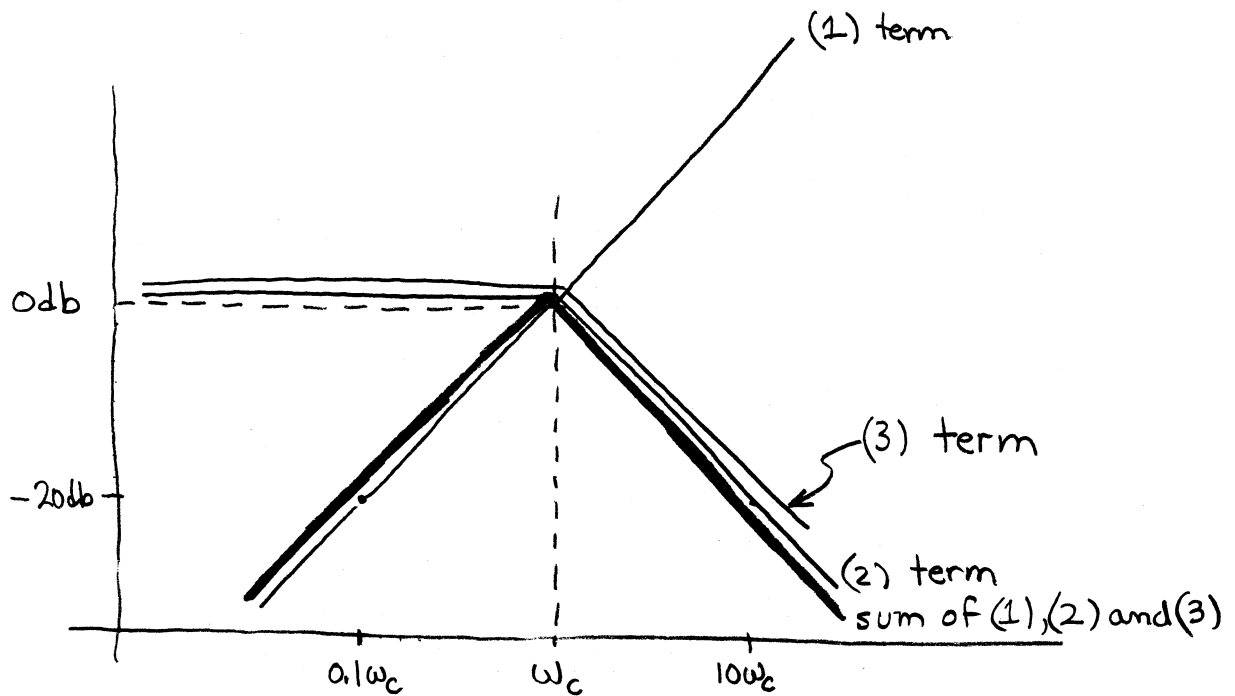
(2) I can neglect the $+1$ term in the denominator.
As we will see this will only change the shape a little
for $\omega \rightarrow 0$

$$T(j\omega) \cong \frac{j\omega(1k)c}{[1 + j\omega(1k)c][1 + j\omega(1k)c]}$$

Taking logs

$$20 \log_{10} |T(j\omega)| \cong 20 \log_{10} |\omega(1k)c| - 20 \log_{10} |1 + j\omega(1k)c| \\ - 20 \log_{10} |1 + j\omega(1k)c|$$

Note: I could have combined the last two terms but do not since they are easier to plot this way.



(1) Start with the $20 \log_{10} |\omega(1k)c|$ term

This term is 0db when $\omega(1k)c = 1$

$$\omega_c = \frac{1}{(1000)(.0047 \times 10^{-6})} = 2.13 \times 10^5 \frac{\text{rad}}{\text{sec}}$$

[in terms of laboratory $f_c \approx \frac{\omega}{2\pi} = 33862 \text{ Hz}$

(2) Plot the second term $-20 \log_{10} |1 + j\omega(1k)c|$

substituting $c = 0.0047 \mu\text{f}$

$$-20 \log_{10} |1 + j\omega(4.7 \times 10^{-6})|$$

$$\text{This term has } \omega_c = \frac{1}{4.7 \times 10^{-6}} = 2.13 \times 10^5 \frac{\text{rad}}{\text{sec}}$$

It is for $\omega \ll \omega_c$ $-20 \log_{10}(1) \rightarrow 0$

$$\omega \gg \omega_c \quad -20 \log_{10} |\omega(4.7 \times 10^{-6})|$$

So it does nothing (0db) for $\omega < \omega_c$ and decreases at -20 dB/decade for $\omega > \omega_c$.

This cancels out the (1) term for $\omega > \omega_c$

(3) The third term repeats the second term

(4) Adding together gives a band-pass characteristic centered at $\omega = 2.13 \times 10^5 \text{ rad/sec}$